

Product of distributions in $BV(\mathbb{R})$.

Daniele Tampieri

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Abstract

In this paper it will be shown how the description of the *electrical regime* of a simple electric circuit leads to precise definition of some product of distributions in $BV(\mathbb{R})$, the space of functions of bounded variation whose domain is \mathbb{R} . Precisely, it became possible to define the product $H(t) \cdot \delta(t)$ as a *Radon measure*.

1 Description of the electric circuit under analysis and statement of the related mathematical problem

In this paper the simple electric circuit (a elementary RC charge circuit) shown in figure 1 below will be analyzed from the point of view of BV functions, in order to show how a simple structure like this give rise to a well defined and meaningful product of distributions. The element labelled V_S is a *constant*

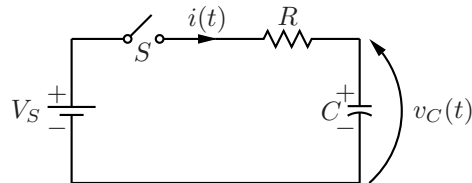


Figure 1: RC charge circuit schematic

voltage generator (practically a high-stability battery) with output voltage V_S , S is a *ideal switch* (i.e. a switch with *negligible turn-on and turn-off times*), while R and C respectively are a *linear resistor* and a *linear capacitor*. Suppose now that the switch S closes for $t = 0$: then the voltage V_S is applied *instantaneously* to the branch made of the series connection of resistor R and capacitor C : A current $i(t) = i_C(t)$ starts to flow through resistor R and charges capacitor C : the voltage $v_C(t)$ across this circuit element rises until the difference $V_S - v_C(t)$ reaches 0. The description of the electrical regime in the network is then given by the *Kirchhoff voltage law*

$$V_S - Ri_C(t) - v_C(t) = V_S - RC \frac{dv_C(t)}{dt} - v_C(t) = 0 \quad t \geq 0$$

together with the knowledge of the voltage across (and therefore the stored charge) the capacitor C at the time $t < 0$

$$v_C(t)|_{t=0^-} = v_0$$

Assuming *zero charge on C* , the electrical regime is therefore described by the following non-homogeneous Cauchy problem

$$\begin{cases} RC \frac{dv_C(t)}{dt} - v_C(t) = V_s H(t) \\ v_C(t)|_{t=0^-} = 0 \end{cases} \quad (1)$$

where

$$H(t) = \begin{cases} 1 & t \geq 0 \\ 0 & t < 0 \end{cases}$$

is the ordinary *Heaviside function*. The solution of this problem has the following form

$$v_C(t) = \begin{cases} V_s \left(1 - e^{-\frac{t}{RC}}\right) & t \geq 0 \\ 0 & t < 0 \end{cases} \quad (2)$$

and consequently

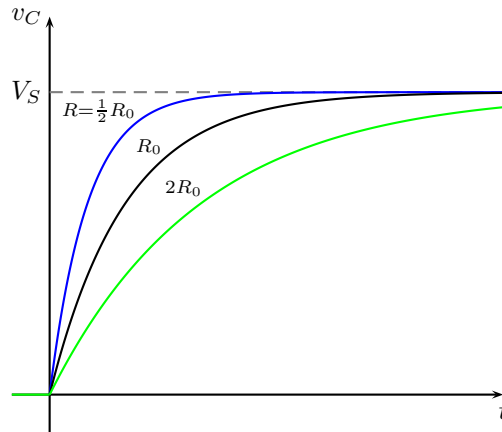


Figure 2: Plot of $v_C(t)$ for different values of R

$$i_C(t) = C \frac{dv_C(t)}{dt} = \begin{cases} \frac{V_s}{R} e^{-\frac{t}{RC}} & t \geq 0 \\ 0 & t < 0 \end{cases} \quad (3)$$

Known the time behavior of the voltage $v_C(t)$, it is sometimes necessary to know how much *energy* \mathcal{E}_C is stored in the capacitor C at the end of the charging process, i.e. for $t = +\infty$. This means that it is necessary to evaluate the following integral

$$\mathcal{E}_C = \int_{-\infty}^{+\infty} v_C(t) i_C(t) dt = C \int_{-\infty}^{+\infty} v_C(t) dv_C(t) = C \int_{\mathbb{R}} v_C(t) dv_C(t) \quad (4)$$

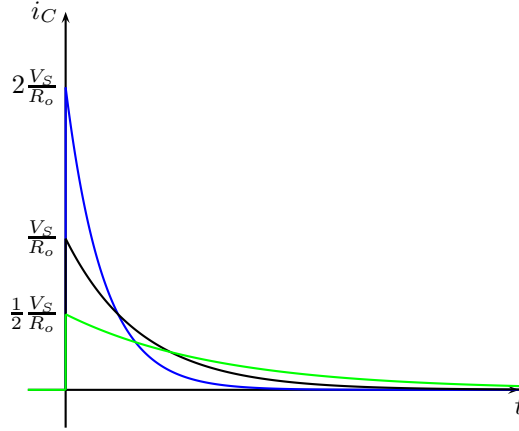


Figure 3: Plot of $i_C(t)$ for different values of R

It is exactly this quantity that, apart from having a precise meaning as a physical entity, has also a precise meaning in the space of BV functions as a *product of distributions*.

2 Analysis of the energy integral \mathcal{E}_C in the space $BV(\mathbb{R})$

The works [3] and [4] are the principal references used in the following text: the Author has preferred their approach even if they deal extensively only with BV functions of *several variables*, since they contain a proof of the *composition formula* for BV functions which is needed to define the product of distribution implied by the structure of the energy integral \mathcal{E}_C .

First of all, a definition of what a BV function is is needed (see pages 158-161 of reference [4]): functions of bounded variations are functions whose *first derivative in the sense of distributions* (a classical reference is [5], page 34) is bounded.

Definition 1 *A real-variable function u is said to be of bounded variation and write $u \in BV(\mathbb{R})$ if and only if there is a Radon measure $du \in \mathcal{M}(\mathbb{R})$ such that*

$$\int_{\mathbb{R}} u(t) \dot{\varphi}(t) dt = - \int_{\mathbb{R}} \varphi(t) du(t)$$

for all functions $\varphi \in C_0^1(\mathbb{R})$ (obviously $\dot{\varphi} \equiv \frac{d\varphi}{dt}$).

Since

$$\int_{\mathbb{R}} v_C(t) \dot{\varphi}(t) dt = -\frac{1}{C} \int_{\mathbb{R}} \varphi(t) i_C(t) dt = -\frac{1}{C} \int_0^{+\infty} \varphi(t) i_C(t) dt$$

the solution (2) of problem (1) is a *function of bounded variation for every value of the resistance R from zero to $+\infty$* . To see this, let $\text{supp}\varphi(t) = [a, b] \in \mathbb{R}$ be

the support of the function $\varphi(t)$: then it is easy to see that

$$\int_{\mathbb{R}} v_C(t) \dot{\varphi}(t) dt = \begin{cases} 0 &]0, +\infty[\cap]a, b] = \emptyset \\ -\frac{1}{C} \int_a^b \varphi(t) i_C(t) dt & a > 0 \\ -\frac{1}{C} \int_0^b \varphi(t) i_C(t) dt & a \leq 0 \end{cases}$$

When $R > 0$, by using the *Mean-Value theorem* (see for example reference [2], pages 689) it is possible to estimate those integrals:

$$\begin{aligned} \int_{\mathbb{R}} v_C(t) \dot{\varphi}(t) dt &= \varphi(\bar{t}) \int_a^b -\frac{V_S}{RC} e^{-\frac{t}{RC}} dt = \\ &= \varphi(\bar{t}) \left[V_S e^{-\frac{t}{RC}} \right]_a^b = V_S \varphi(\bar{t}) \left(e^{-\frac{b}{RC}} - e^{-\frac{a}{RC}} \right) \end{aligned} \quad (5)$$

for $a > 0$ with $\bar{t} \in [a, b]$ a proper fixed point, and

$$\begin{aligned} \int_{\mathbb{R}} v_C(t) \dot{\varphi}(t) dt &= \varphi(\bar{t}) \int_0^b -\frac{V_S}{RC} e^{-\frac{t}{RC}} dt = \\ &= \varphi(\bar{t}) \left[V_S e^{-\frac{t}{RC}} \right]_0^b = V_S \varphi(\bar{t}) \left(e^{-\frac{b}{RC}} - 1 \right) \end{aligned} \quad (6)$$

for $a \leq 0$ with $\bar{t} \in [0, b]$ a proper fixed point.

When $R = 0$, remembering that $BV(\mathbb{R})$ is a Banach space (see references [4] pages 158-152), every Cauchy sequence in this space has a limit belonging to it, therefore it suffices to evaluate *the limit* of the above integral for R going to 0. The integral (5), where $0 \notin [a, b]$, goes to 0 for R going to 0 since $V_S \varphi(\bar{t})$ is bounded. Evaluation of the integral (6), where $0 \in [a, b]$, needs a little more work: by using the Mean-Value theorem and choosing a proper *time value* $k\sqrt{RC} \in]\varepsilon, b - \varepsilon[$ for $\varepsilon > 0$, this integral can be decomposed into two integrals on the intervals $[0, k\sqrt{RC}]$ and $[k\sqrt{RC}, b]$. Therefore

$$\begin{aligned} \int_{\mathbb{R}} v_C(t) \dot{\varphi}(t) dt &= \varphi(\bar{t}_1) \int_0^{k\sqrt{RC}} -\frac{V_S}{RC} e^{-\frac{t}{RC}} dt + \varphi(\bar{t}_2) \int_{k\sqrt{RC}}^b -\frac{V_S}{RC} e^{-\frac{t}{RC}} dt = \\ &= \varphi(\bar{t}_1) \left[V_S e^{-\frac{t}{RC}} \right]_0^{k\sqrt{RC}} + \varphi(\bar{t}_2) \left[V_S e^{-\frac{t}{RC}} \right]_{k\sqrt{RC}}^b = \\ &= V_S \left[\varphi(\bar{t}_1) \left(e^{-\frac{k}{\sqrt{RC}}} - 1 \right) + \varphi(\bar{t}_2) \left(e^{-\frac{b}{RC}} - e^{-\frac{k}{\sqrt{RC}}} \right) \right] \end{aligned}$$

where $\varphi(\bar{t}_1) \in [0, k\sqrt{RC}]$ and $\varphi(\bar{t}_2) \in [k\sqrt{RC}, 0]$: since

$$\lim_{R \rightarrow 0} \varphi(\bar{t}_1) e^{-\frac{k}{\sqrt{RC}}} = \lim_{R \rightarrow 0} \varphi(\bar{t}_2) e^{-\frac{k}{\sqrt{RC}}} = \lim_{R \rightarrow 0} \varphi(\bar{t}_2) e^{-\frac{b}{RC}} = 0$$

and since $\varphi(\bar{t}_1)$ tends to $\varphi(0)$ because $[0, k\sqrt{RC}]$ tends $\{0\}$ as R tends to 0, the limit of the given integral is

$$\int_{\mathbb{R}} v_C(t) \dot{\varphi}(t) dt = -V_S \varphi(0)$$

therefore

$$dv_C(t) = V_S \delta(t) dt \quad (7)$$

i.e. the limit of the weak derivative of the voltage $v_C(t)$ as R tends to zero is the Dirac measure multiplied by a constant term V_S (see reference [5] page 36).

Now it is possible to deal with the basic problem stated in the introduction: first of all it is necessary to note that if $u(t)$ is a BV function of one variable the following limits exists and are finite

- $u_+(t) = \lim_{s \rightarrow t^+} u(s)$, the right limit of the function u at the point $t \in \mathbb{R}$
- $u_-(t) = \lim_{s \rightarrow t^-} u(s)$, the left limit of the function u at the point $t \in \mathbb{R}$

For a proof of this assertion, see references [3], pages 237-243, and [4], pages 161-164: this fact implies that BV functions of one variable have only jump-type singularities and also that the following definition is well-posed.

Definition 2 Let $f : \mathbb{R} \rightarrow \mathbb{R}$ a continuously differentiable function. Given a function $u \in BV(\mathbb{R})$ the averaged superposition is defined as

$$\bar{f}(u(t)) = \int_0^1 f(u_+(t)s + u_-(t)(1-s)) ds \quad (8)$$

Acknowledged the concept of averaged superposition, it is possible to formulate the following

Theorem 1 Let $f : \mathbb{R}^p \rightarrow \mathbb{R}$ be a continuously differentiable function. Given functions $\mathbf{u}(t) = (u_1(t), \dots, u_p(t)) \in (BV(\mathbb{R}))^p$ and let the averaged superposition $\partial \bar{f}(\mathbf{u}(t)) / \partial u_k$ be locally summable respect to the Radon measure $du_k(t)/dt$ for $k = 1, \dots, p$. Then $f(\mathbf{u}(t)) \in BV(\mathbb{R})$ and the formula

$$\frac{df(\mathbf{u}(t))}{dt} = \sum_{k=1}^{k=p} \frac{\partial \bar{f}(\mathbf{u}(t))}{\partial u_k} \frac{du_k(t)}{dt} \quad (9)$$

holds

For a proof of this important theorem in the multi-dimensional setting see reference [3], pages 248-250. Then, since $v_C(t) = V_S H(t)$ and $\frac{dv_C(t)}{dt} = V_S \delta(t)$ in the limiting case $R = 0$, considering the function $f(u(t)) = u^2(t)$ formula 9 reads as follows

$$\frac{d}{dt} v_C^2(t) = 2\bar{v}_C(t) \frac{dv_C(t)}{dt} = 2 \cdot V_S \bar{H}(t) \cdot V_S \delta(t) = 2V_S^2 \bar{H}(t) \cdot \delta(t) \quad (10)$$

having defined

$$\bar{H}(t) = \frac{1}{2} [H_+(t) + H_-(t)] = \begin{cases} 0 & t < 0 \\ 1/2 & t = 0 \\ 1 & t > 0 \end{cases}$$

and since

$$\begin{aligned} \frac{\partial \bar{f}(v_C(t))}{\partial u} &= 2\bar{v}_C(t) = 2 \int_0^1 (v_{C+}(t)s + v_{C-}(t)(1-s)) ds = \\ &= 2 \left(v_{C+}(t) \left[\frac{1}{2} s^2 \right]_0^1 - v_{C-}(t) \left[\frac{1}{2} (1-s)^2 \right]_0^1 \right) = \\ &= [v_{C+}(t) + v_{C-}(t)] \end{aligned}$$

It is now possible to give the following definition:

$$V_S H(t) \cdot V_S \delta(t) \stackrel{\text{def}}{=} \frac{1}{2} \frac{d}{dt} v_C^2(t) = \frac{V_S^2}{2} \cdot \delta(t) \iff H(t) \cdot \delta(t) = \frac{1}{2} \delta(t) \quad (11)$$

this product of distribution is well-defined and has also a very simple physical mean: since for each $R \geq 0$ the following equality holds true

$$\mathcal{E}_C = C \int_{\mathbb{R}} v_C(t) dv_C(t) = C \int_0^{+\infty} v_C(t) dv_C(t) = \frac{1}{2} C V_S^2$$

then equation (11) represents the power sinked by (or the power sourced by the generator V_S to) the capacitor C in the limiting case of zero resistance of the charging circuit. Note that it is possible to give the same definition of this product even by using only the *Leibnitz formula* (proved in reference [4], pages 189-191) i.e. formula (9) for the function $f(u_1(t), u_2(t)) = v_C(t)u(t)$

$$\frac{d}{dt} v_C(t)u(t) = \bar{u}(t) \frac{dv_C(t)}{dt} + \bar{v}_C(t) \frac{du(t)}{dt}$$

and this is a consequence of the coherent structure of the calculus of *BV* functions.

As a last observation for this section, it is worth noting that the *total energy* \mathcal{E}_S sourced by the generator V_S has the following expression

$$\mathcal{E}_S = \int_{-\infty}^{+\infty} V_S(t) i_C(t) dt = C \int_{\mathbb{R}} V_S(t) dv_C(t) = C V_S \int_0^{+\infty} dv_C(t) = C V_S^2$$

whenever $R > 0$ or $R = 0$. This means that *apparently, half of the total energy sourced by the generator V_S has disappeared during the charging process in the limiting case $R = 0$* : when $R > 0$ the “lost” energy is dissipated in the resistor R as Joule’s Law predicts.

3 Conclusions

The consequences of the analysis done in the preceding sections are essentially two:

1. A formula for the distributional product of the Heaviside function and the Dirac distribution was given and it was shown that this formula has a very simple physical meaning. It is worth to notice that the formula was obtained by means of “classical analysis” in the function space $BV(\mathbb{R})$, without recourse to the methods of the theory of distributions.
2. It was shown that the given formula implies that *the model physical system violates the energy balance principle* for the model physical system (a simple *RC* charge circuit). This is only due to the fact that for the limiting case $R = 0$ this model is no more valid: it is necessary to use the full *Maxwell system* to determine the electrical regime of this system, and therefore to know its *geometrical shape*. Precisely, the *energy balance principle* for a electromagnetic system looks like

$$\frac{d}{dt} \mathcal{E}_{\text{tot}} = \frac{d}{dt} \mathcal{E}_{\text{cond}} + \frac{d}{dt} \mathcal{E}_{\text{rad}} \quad (12)$$

where

- (a) $\frac{d}{dt}\mathcal{E}_{\text{tot}}$ is the *total power* generated and dissipated by the system,
- (b) $\frac{d}{dt}\mathcal{E}_{\text{cond}}$ is the power generated and/or dissipated by physical processes involving conduction or more generally *movements of charges in solids*,
- (c) $\frac{d}{dt}\mathcal{E}_{\text{rad}}$ is the *radiated power* i.e. the *flux of the Poynting vector \mathbf{S}* (see for example [1]) flowing across a surface Σ of arbitrary shape and given *normal vector $\hat{\mathbf{n}}$* , enclosing the given physical system

$$\frac{d}{dt}\mathcal{E}_{\text{rad}} = \int_{\Sigma} \mathbf{S} \cdot \hat{\mathbf{n}} d\sigma$$

This second term of equation (12) is *barely neglected* when using a *circuital* i.e. *quasi-stationary description* of the system, and this is no more possible in the limiting case $R = 0$.

The author asks if the second occurrence described above is to be found in every model system where a product of distributions arises which violates a *energy balance principle*, i.e if there is a more complex physical description giving rise to an *integral balance law* where this product appears together with a neglected quantity, restoring equality.

References

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